40G/100G Transmitter Compliance Testing

40G TESTING RECOMMENDATIONS

The ITU-T G.959.1 (line side) and IEEE 802.3ba-2010 (client side) standards define the requirements for transmitter pulse shape characteristics such as rise time, fall time, pulse overshoot, pulse undershoot and ringing. These parameters are specified by a mask in the transmitter eye diagram. A good transmitter eye diagram does not cross any of the hatched or grey areas. The recommended masks for non-return-to-zero (NRZ) 40G, return-to-zero (RZ) 40G/100G and the mask parameters are shown below.

These masks allow control of several potential transmitter impairments:

- > Timing jitter > ISI interference
- > RF levels > Modulator bias
- For DP-QPSK and other phase-modulated signals, it is recommended to use a constellation diagram (see flip side of poster) in conjunction with the eye diagrams to fully characterize a transmitter.

40G NRZ MASK FOR LINE SIDE **AS RECOMMENDED BY ITU-T**



40G NRZ MASK FOR LINE SIDE AS RECOMMENDED BY ITU-T



	NRZ 40G		
X4-X1	N/A		
X1-X2	N/A		
X3-X2	0.2		
Y1	0.25		
Y2	0.75		
Y3	0.25		
Y4	0.25		

ffs = for further study

Source: Recommendation ITU-T G.959.1 (11/2009), Optical Transport Network Physical Layer Interfaces.



	40GBASE-SR4 100GBASE-SR10	40GBASE-LR4 100GBASE-LR4/ER4
X1	0.23	0.25
X2	0.34	0.4
X3	0.43	0.45
Y1	0.27	0.25
Y2	0.35	0.28
Y3	0.4	0.4

Source: The figure and the table above are reprinted with permission from IEEE Std.802.3ba-2010*, Telecommunications and information exchange between systems—Local and metropolitan area networks—Specific requirements—Part 3: Carrier Sense Multiple Access with Collision Detection (CSMA/CD) Access Method and Physical Layer Specifications*, Copyright 2010*, by IEEE. The IEEE disclaims any responsibility or liability resulting from the placement and use in the described manner.

ERROR VECTOR MAGNITUDE (EVM), BIT ERROR RATE (BER) AND SIGNAL-TO-NOISE RATIO (SNR)



- > The error vector magnitude (EVM) is the comparison of the received signal with the ideal signal taking into consideration both the phase and the magnitude errors.
- > Its value is calculated using the formula:

$$\Xi VM_{i} = \sqrt{\frac{\Delta I_{i}^{2} + \Delta Q_{i}^{2}}{\sqrt{\Delta I^{2} + \Delta Q^{2}}}}$$

where EVMi is the error vector magnitude for point I, Δ I and ΔQ are defined in the figure above; I and Q, and the x and y coordinates of the measured value.

RZ 40G	
fs	
fs	
N/A	
fs	
fs	
fs	
fs	

TIME-RESOLVED ERROR VECTOR MAGNITUDE

Time-resolved EVM is a unique approach developed by EXFO to ease and fasten transmitter compliance testing, enabling mask testing by EVM. This approach is closely related to the familiar concept of eye-diagram mask testing. It consists of displaying the EVM value as a function of time, more specifically as a function of the position in the symbol slot. It uses the same principle as the eye diagram, which consists of combining several periods into a single diagram.

Time-resolved EVM combines the strengths of both EVM and eye-diagram analysis and provides complete transmitters' characterization for multilevel signaling in a single measurement.





Symbol slot



Symbol slot

Symbol slot

> This figure assumes that the noise distribution is Gaussian and that the number of symbols in the measured constellation is much greater than the number of unique modulation symbols.

Impairments clearly reduce margins out do not affect the EVM at center

40G/100G Transmitter **Analysis Reference Poster**







EXFO'S PSO-200 MODULATION ANALYZER

- > Supports data rates of 40 Gbit/s, 100 Gbit/s, 400 Gbit/s, 1 Tbit/s and beyond
- > For NRZ, RZ, DPSK, DQPSK, QPSK, 16-QAM
- > Single- or dual-polarization transmission
- > Distortion-free signal recovery

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EXAMPLE OF AVERAGED QPSK TIME-RESOLVED EVM



EXAMPLE OF AVERAGED QPSK TIME-RESOLVED EVM WITH 4 PS SKEW



Advanced Modulation Schemes and Impairments in 40 Gbit/s and 100 Gbit/s Networks

CONSTELLATION DIAGRAM

A constellation diagram is a representation of a signal modulated by a digital modulation scheme (phase and/or amplitude). In other words, it shows the possible symbols that can be selected by a given modulation format as points in the complex plane.



Example of a quadrature phase-shift keying (QPSK) constellation diagram

I = In-phase axis or real part of the signal

Q = Quadrature axis or imaginary part of the signal

MODULATION SCHEMES

ON/OFF KEYING (OOK) NON-RETURN-TO-ZERO (NRZ)	Amplitude modulation Presence of signal = '1' Absence of signal = '0' One bit encoding	Amplitude $ \begin{array}{c c} & & \\ \hline & & \\ & $
ON/OFF KEYING (OOK) RETURN-TO-ZERO (RZ)	Amplitude modulation Presence of signal = '1' Absence of signal = '0' One bit encoding	Amplitude $\begin{array}{c c} & & & \\ \hline \\ 1 \\ 1 \\ 0 \\ 0 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1$
BINARY PHASE-SHIFT KEYING (BPSK)	Phase modulation Two possible values E.g.: - Phase of $0^{\circ} = '1'$ - Phase of $180^{\circ} = '0'$ One bit encoding	Amplitude 0 1 1
DIFFERENTIAL PHASE-SHIFT KEYING (DPSK)	Same as BPSK, but data is encoded in the phase difference between adjacent symbols E.g.: - Phase of $0^\circ = '0'$ - Phase of $180^\circ = '1'$	Amplitude
QUADRATURE PHASE-SHIFT KEYING (QPSK)	Phase modulation Quadrature phase shifts are separated by 90° (e.g., 45°, 135°, 225°, 315°) Two bits encoding	Amplitude 01 11 01
DIFFERENTIAL QUADRATURE PHASE-SHIFT KEYING (DQPSK)	Same as QPSK, but data is encoded in the phase difference between adjacent symbols Phase shift = 0°, 90°, 180° or 270° (each phase shift depends on encoding scheme) Two bits encoding	Amplitude
16 QUADRATURE AMPLITUDE MODULATION (16-QAM)	Both waves (I and Q) show a phase difference of 90° and their amplitude has four discrete levels Four bits encoding	Amplitude

BLOCK DIAGRAM



POOR SIGNAL-TO-NOISE **RATIO TRANSMITTER**



Clouded constellation and eye diagrams are typically of poor SNR due to an instrument limitation.

I/Q QUADRATURE ERROR



A rhombic constellation appears when the I and Q phases do not show a perfect 90° phase shift, which occurs when bias B5 is not optimized.

DETERMINISTIC DATA DEPENDENT JITTER



The I and Q RF drive signals (RF3 and RF4) may contain deterministic jitter originating from driver circuits or SERDES that leads to a delay in the transitions.

CHIRP

The S-shape transitions of the chirp impairment can stem from data modulation or from residual fiber dispersion.

I/Q MODULATOR BIAS ERROR





This impairment, caused by an incorrect bias in the I-branch of the I/Q modulator (bias B1), results in an overshoot in the I direction and an undershoot in the Q direction.

RANDOM DATA CLOCK JITTER



An equal delay in the I and Q phases due to clock jitter (RF3 and RF4 drive signals) leads to an impairment that is only visible in the eye diagram.

Assessing

Next-Gen Networks



EXFO